



# Model Analysis of Transformers

Advanced training course C

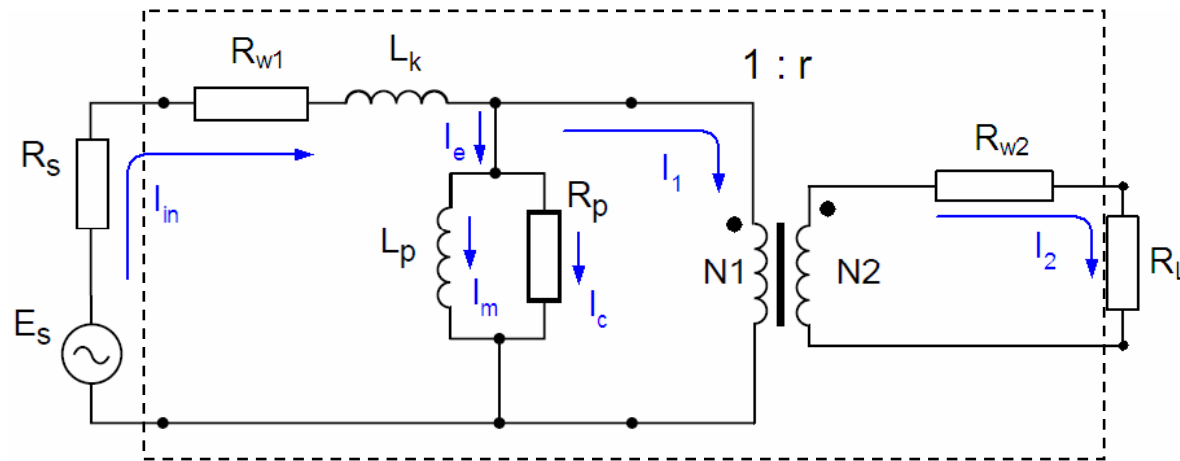
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## • Equivalent circuit of a transformer



$E_s$ : Voltage source

$R_s$ : Source impedance

$R_{w1}$ : Primary winding resistance

$L_k$ : Total leakage inductance

$L_p$ : Primary self inductance in parallel mode

$R_p$ : Core loss resistance in parallel mode

$R_{w2}$ : Secondary winding resistance

$R_L$ : Load resistance

$N1$ : Primary winding turns

$N2$ : Secondary winding turns

$r$ : Turn ratio

$I_{in}$ : Input current

$I_1$ : Reflected current from  $I_2$

$I_2$ : Load current

$I_e$ : Exciting current

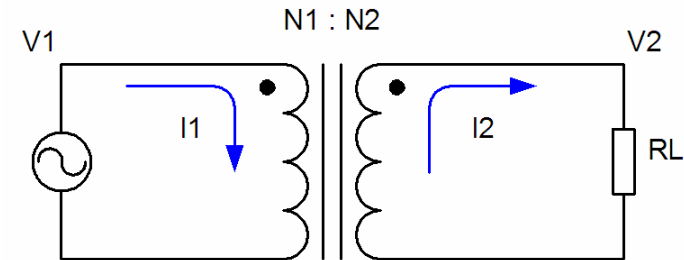
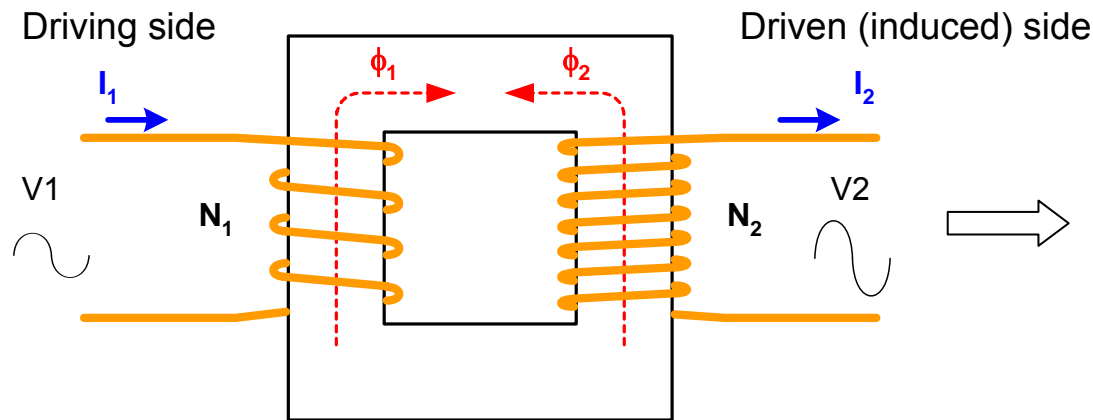
$I_m$ : Magnetizing current

$I_c$ : Core loss current





## • Ideal transformer theory



### Polarity dot rule of a transformer circuit:

*When  $I_1$  flows into the primary dot side, then  $I_2$  will flow out from the secondary dot side. Moreover,  $I_1$  follows the waveform of  $I_2$  and with the same phase.*

①  $\frac{V_1}{V_2} = \frac{N_1}{N_2}$   $\Rightarrow$  100% coupling, no flux leakage and stray capacitance

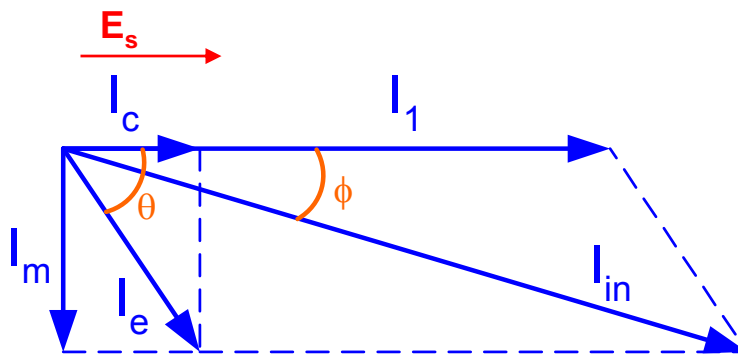
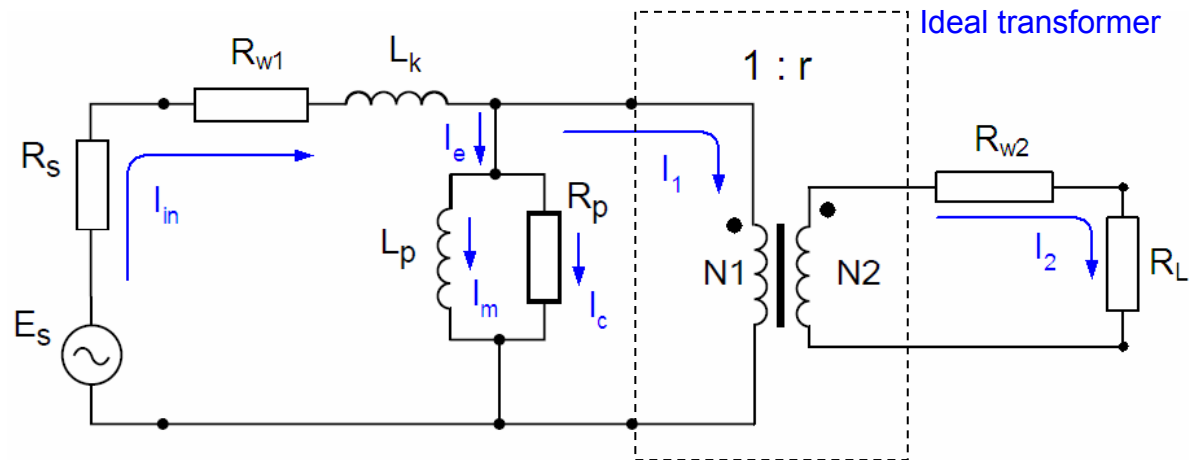
②  $N_1 \times I_1 = N_2 \times I_2$   $\Rightarrow$  Flux  $\phi_1$  driven by magneto-motive force  $N_1 \times I_1$  equals flux  $\phi_2$  driven by induced magneto-motive force  $N_2 \times I_2$ .

③ No winding loss and no core loss;  $L_p$  &  $R_p \rightarrow \infty$





## • Vector diagram of transformer currents



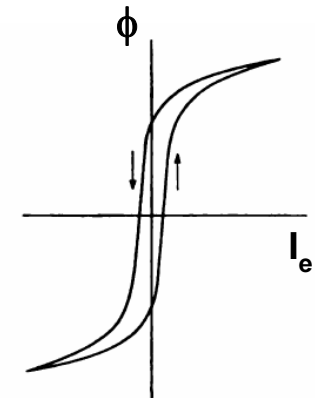
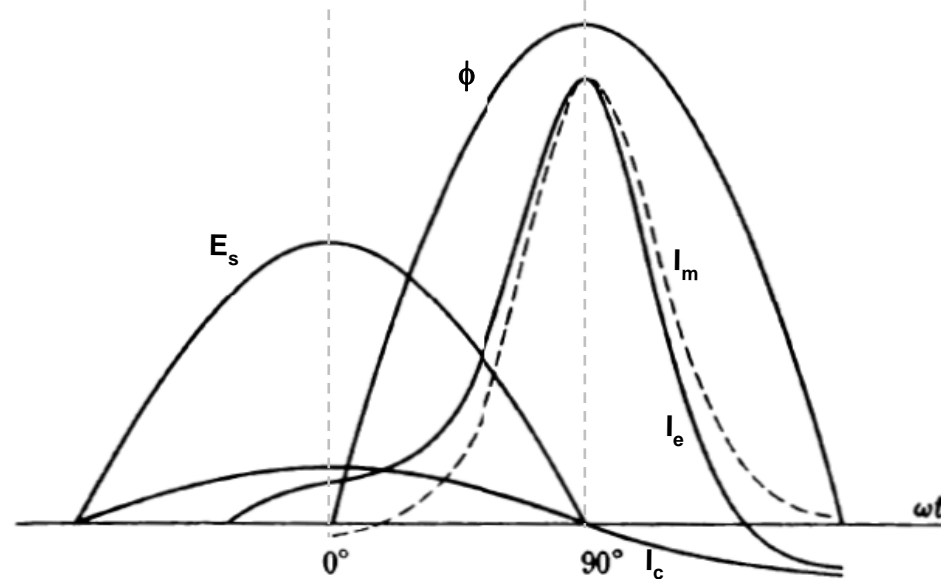
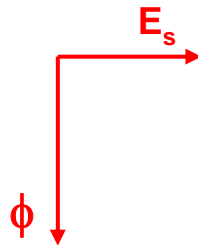
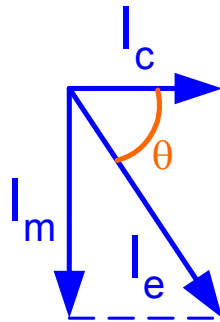
Vector set,  $I_e = I_m + I_c$ , is decided by the core material and the primary inductance. Ideally it's nothing to do with the reflected current  $I_1 = r \times I_2$ .

This vector set can be worked out by measuring the input current  $I_{in}$  and PF ( $\cos \theta$ ) at open load condition.





## • Waveform analysis of exciting current $I_e$ – practical case



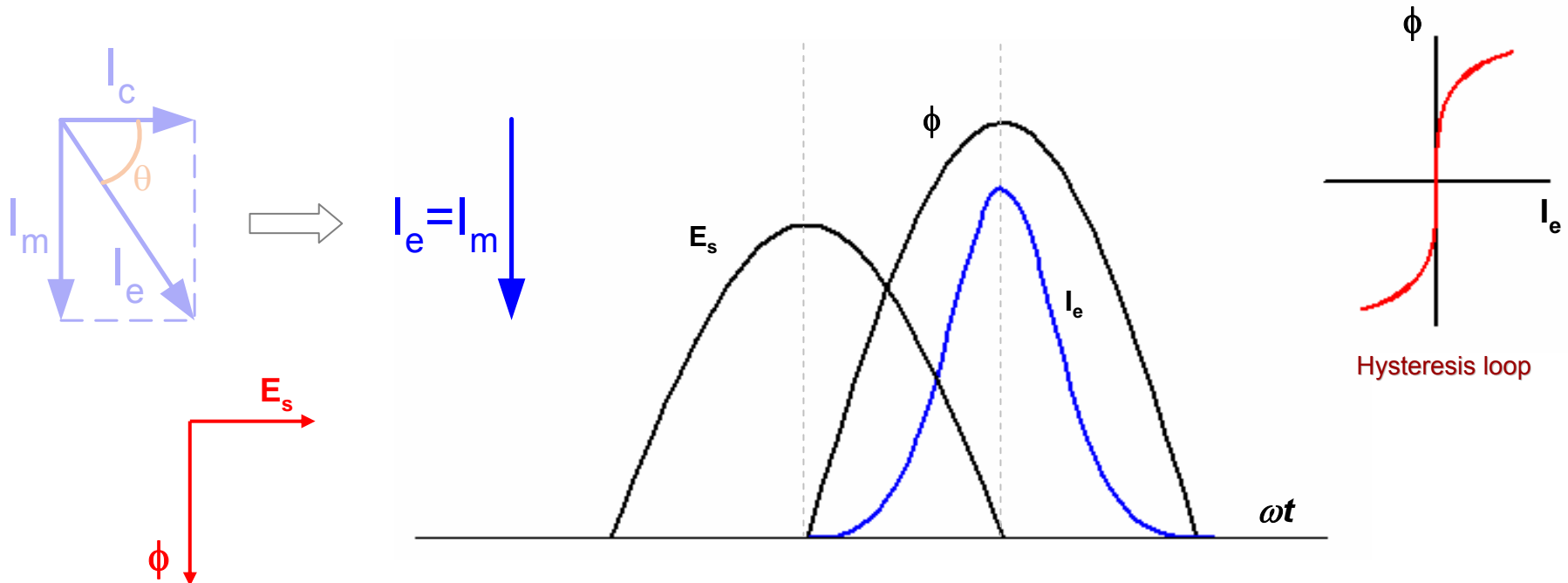
Hysteresis loop

If the applied voltage  $E_s$  is sinusoidal, the waveform of induced flux  $\phi$  also is very nearly sinusoidal. However, the resulted exciting current will go non-sinusoidally with lots of harmonic components due to the hysteresis loop of the core material. The core loss current  $I_c$  is in phase with  $E_s$  and goes sinusoidally. **Thus, the core loss  $P_c$  can be calculated by  $E_s \cdot I_c$ .**





- Waveform analysis of exciting current – if ignoring core loss effect

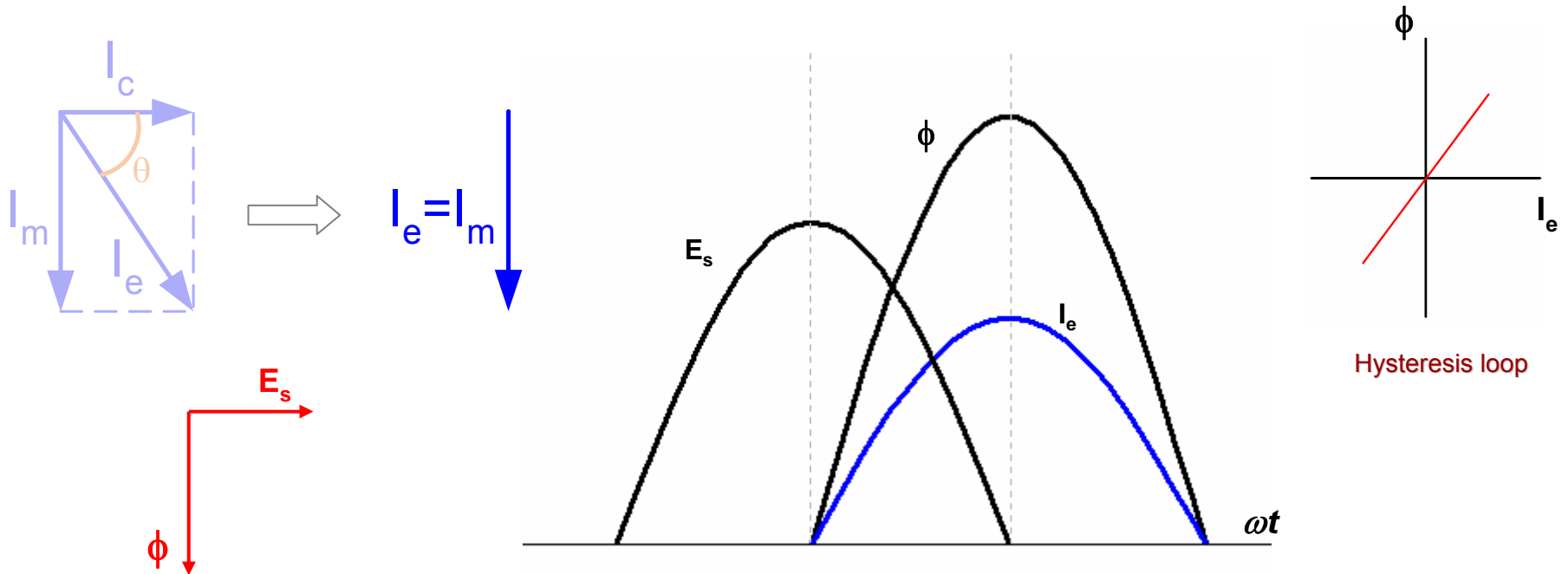


If the applied voltage  $E_s$  is sinusoidal, the waveform of induced flux  $\phi$  also is very nearly sinusoidal. The resulted exciting current still will go non-sinusoidally with some harmonic components even if the hysteresis loop of the core material has no core loss.





- Waveform analysis of exciting current – if B-H curve is linear

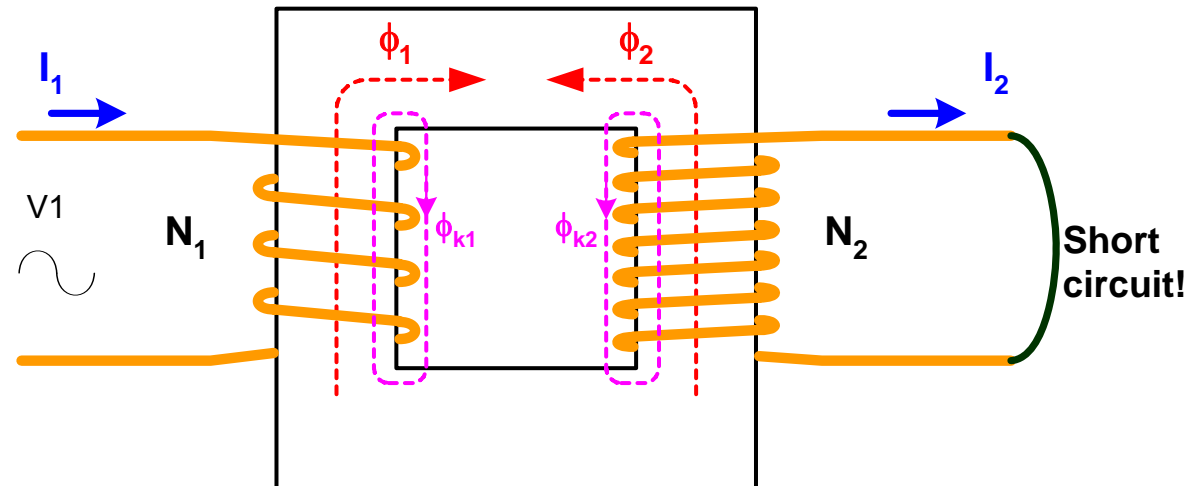


If the applied voltage  $E_s$  is sinusoidal, the waveform of induced flux  $\phi$  also is very nearly sinusoidal. The resulted exciting current will go sinusoidally if only the B-H curve of the core material is linear. **This is typically the design case of isolation transformers for xDSL modem to reduce the THD (Total Harmonic Distortion) which is caused by the harmonic components of the exciting current.**



## • Leakage inductance $L_k$

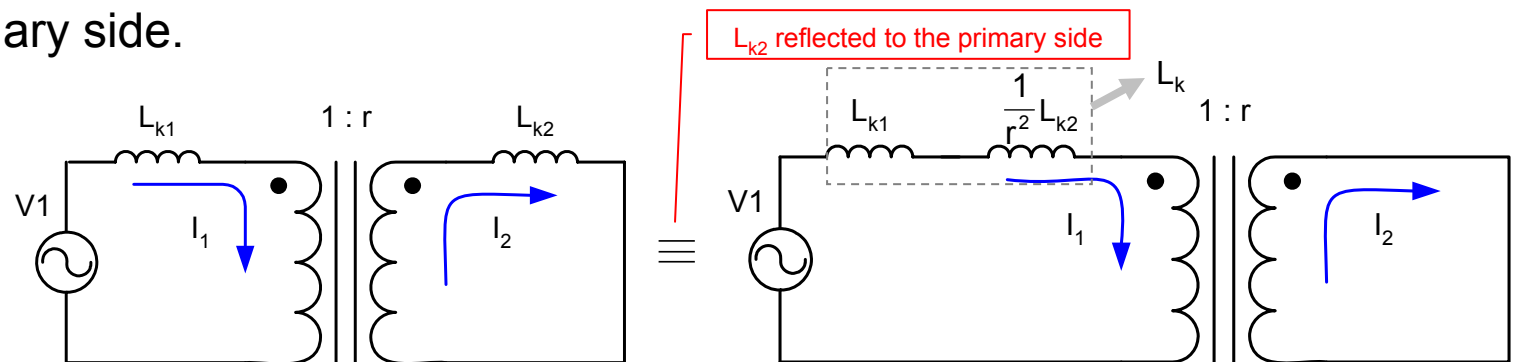
Ideally the flux  $\phi_1$  driven by magneto-motive force  $N_1 \times I_1$  will be cancelled by flux  $\phi_2$  driven by induced magneto-motive force  $N_2 \times I_2$ . However, in practice there exist leakage flux  $\phi_{k1}$  of the primary winding and  $\phi_{k2}$  of the secondary winding. As a result of that, the leakage inductance  $L_k$  can be measured by short circuit at the secondary side.



$$L_{k1} = \frac{N_1 \cdot \Delta \phi_{k1}}{\Delta I_1}$$

$$L_{k2} = \frac{N_2 \cdot \Delta \phi_{k2}}{\Delta I_2}$$

$$L_k = L_{k1} + \frac{1}{r^2} L_{k2}$$







## • The efficiency $\eta$ of power transformers

$$\eta = \frac{P_o}{P_{in}} = \frac{P_o}{P_o + P_t}$$

$$P_t = P_c + P_w = I_c \cdot E_s + (I_{in}^2 \times R_{w1} + I_2^2 \times R_{w2})$$

$$P_o = r E_s \cdot I_2$$

$$I_{in} = (I_c + r I_2) + I_m$$

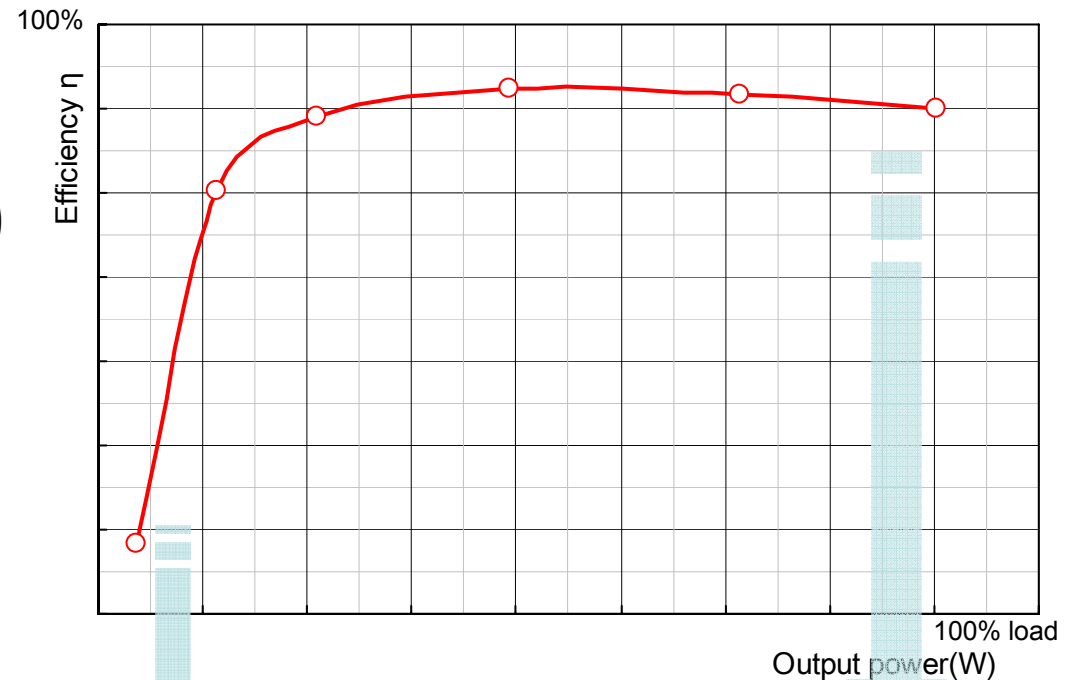
$P_o$ : Output power

$P_{in}$ : Input power

$P_t$ : Total transformer loss

$P_c$ : Core loss

$P_w$ : Winding loss



**Core loss dominates!**

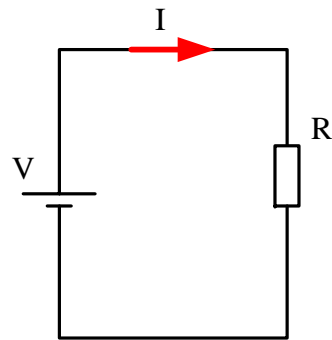
**Winding loss dominates!**



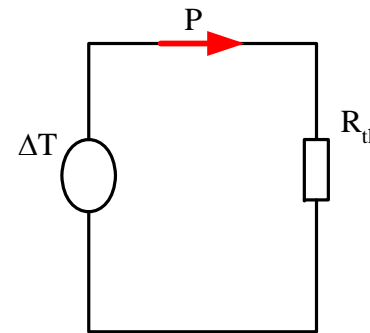
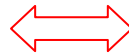


- **Temperature rise estimation of power transformers**

Thermal resistance model in analogy to electric circuit:



Ohm's law:  $V = I \times R$



Fourier's law:  $\Delta T = P \times R_{th}$

, in which  $\Delta T$  is the temperature rise,  $P$  is the heat transfer rate or power loss and  $R_{th}$  is the thermal resistance of the device.

**It should be noted that the heat will in general be transferred to the surroundings by means of conduction, convection and by radiation, and each of these processes may be natural or may be artificially added. As a result of that, the concerned thermal resistance  $R_{th}$  will be dependent. One can refer to the table 9.3 of Snelling's book\* to get  $R_{th}$  for temperature rise estimation.**

\* *E.C. Snelling, Soft Ferrites – Properties and Applications, 2nd Edition, Butterworths, London (1988)*





Thank you for listening!

